
Innovative Approaches to Antenna Characterization: A Compressive Sensing Framework

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1 Magnitude Failure Factor and Phase Shift values impact on the Near Field Error

1.1 Test performed using the parameters : $M = 25$, $F^{(s)} = 7$ and $P^{(s)} = 5$

By using a number of :

- simulated configurations to build the over-complete basis :
 - number of failure factors, $F^{(s)} = 7$;
 - number of phase shifts, $P^{(s)} = 5$;
- measurement points, $M = 25$;

the objective of this section is to verify if the near-field error has a relation with the value of the failure (either in magnitude or in phase) or if it is completely insensitive to the failure intensity.

1.1.1 AUT with only a magnitude failure affecting the 3rd row ($\nu^{(3)} = [0, 0.15, 0.45, 0.65, 1]$); incremented failure ranges to build the over-complete basis ($\nu^{(s)} \in [0.0, 1.0]$, $F^{(s)} = 7$, $\gamma^{(s)} \in [-\pi, \pi]$, $P^{(s)} = 5$) and M = 25

Parameters

Gold Antenna (Without Defects)

- Geometry : Planar array of microstrip patches on the (x, y) plane;
- Working Frequency : $f = 3.6 [GHz]$ ($\lambda = 83.27 \times 10^{-3} [m]$ in free space);
- Substrate (PEC-backed) :
 - Dimensions : infinite;
 - Relative Permittivity : $\varepsilon_{r,sub} = 4.7$;
 - Loss Tangent : $\tan \delta_{sub} = 0.014$;
 - Thickness : $h_{sub} = 0.019 [\lambda]$ (1.6 [mm]);
- Microstrip patches :
 - Dimensions : $l_x \approx 0.22 [\lambda]$ (18.16 [mm]), $l_y \approx 0.33 [\lambda]$ (27.25 [mm]);
 - Feeding : pin-fed;
- Spacing between elements : $d_x = d_y = \frac{\lambda}{2}$;
- Number of elements in each row : $N_x = 6$;
- Number of elements in each column : $N_y = 10$;
- Total number of elements : $N = (N_x \times N_y) = 60$;
- Total size of the antenna : $L_x = 5 [\lambda]$, $L_y = 9 [\lambda]$;
- Element excitations : $w_n^{(s)} = 1.0 + j0.0$, $n = 1, \dots, N^{(s)}$, $s = 1, \dots, S$;

Antenna Under Test (AUT - With Defects)

1. Failures of the excitation magnitude of the 3rd row;
 - Magnitude failure of the elements in the 3rd row ($s = 3$) : $\nu^{(3)} = [0, 0.15, 0.45, 0.65, 1]$;

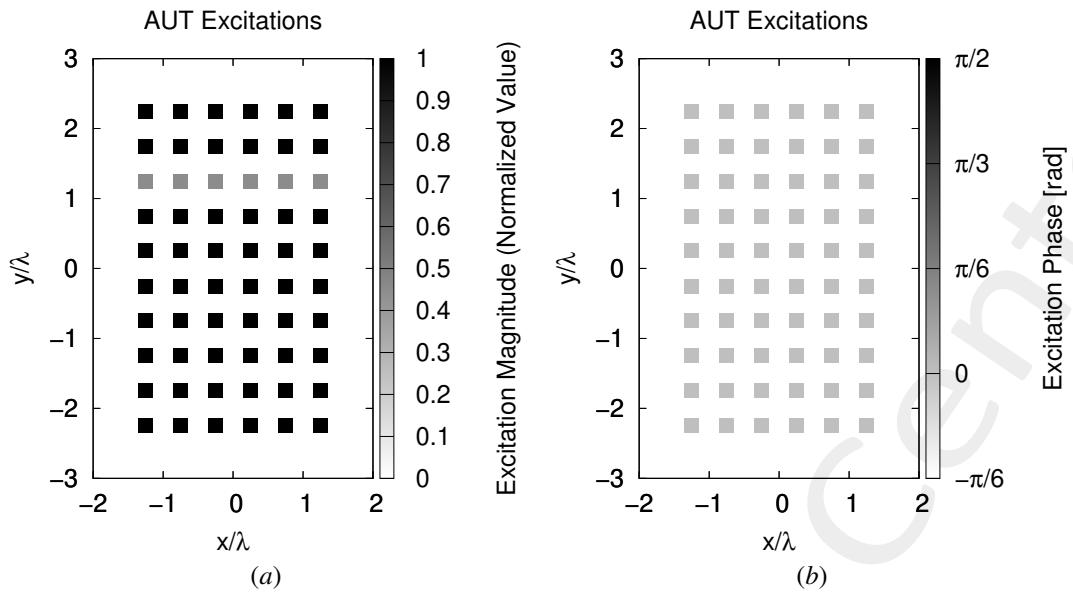


Figure 1: (a) Magnitude of the element excitations in the *AUT* (e.g. $\nu^{(3)} = 0.45$), (b) phase of the element excitations in the *AUT*.

Measurement Set-Up

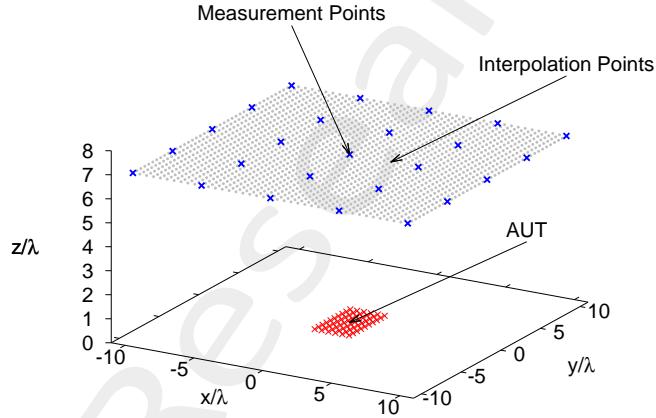


Figure 2: Disposition of the interpolation points ($T = 1681$) and of the measurement points ($M=25$) in the near-field region of the *AUT*.

- Type of measurements : near-field;
- Height of the measurement region : $H = 7 [\lambda]$;
- Interpolation points :
 - Number of points : $T = 41 \times 41 = 1681$;
 - Coordinates : $x_t \in [-10, 10] [\lambda], y_t \in [-10, 10] [\lambda], z_t = H [\lambda], t = 1, \dots, T$;
 - Interpolation step : $\Delta_{x/y}^{int} = 0.5 [\lambda]$;
- Measurement points :
 - Coordinates : $x_m^{meas} \in [-10, 10] [\lambda], y_m^{meas} \in [-10, 10] [\lambda], z_m^{meas} = H [\lambda], m = 1, \dots, M$;

- Number of points : $M = 25$;
- Measurement step : $\Delta_{x/y}^{meas} = 5 [\lambda]$
- Ratio between number of measurements and total number of elements : $(M/N) = 0.42$;

Measurement-by-Design Technique

- Number of generated bases : $B = 20$;
- Bases $b = 1, \dots, 10$: magnitude failures in each row ($s = 1, \dots, 10$)
 - Failure factor of the elements : $\nu^{(s)} = [0.0, 1.0]$, $s = 1, \dots, 10$;
 - Number of simulated failure factors : $F^{(s)} = 7$, $s = 1, \dots, 10$;
- Bases $b = 11, \dots, 20$: phase failures in each row ($s = 1, \dots, 10$)
 - Phase shift of the elements : $\gamma^{(s)} \in [-\pi, \pi] [rad]$, $s = 1, \dots, 10$;
 - Number of simulated phase shifts: $P^{(s)} = 5$, $s = 1, \dots, 10$;
- Threshold on the singular values magnitude (normalized) : $\eta = -40 [dB]$;
- Total number of simulated AUT configurations : $K = S \times (F^{(s)} + P^{(s)}) = 10 \times (7 + 5) = 120$;

Dimension of the Over-Complete Basis

The dimension of the over-complete basis is

$$Q = 40$$

This number is given by the sum of the vectors belonging to the two considered bases:

1. Magnitude failures : $Q_1, \dots, Q_{10} = 2$;
2. Phase failures : $Q_{11}, \dots, Q_{20} = 2$.

Alternative (BCS) MbD parameters

- Toleration factor for BCS solver: $Tolerance = 1 \times 10^{-8}$;
- Initial noise variance for BCS solver: $\eta_0^{opt_1} = 10^{-2}$. This value has been obtained as a result of a calibration procedure;

Original (OMP) MbD parameters

- Max. number of iterations of the OMP algorithm : $I = \{1; 2; 3; \dots; 10\}$;
- Selected iteration to report the results: depends on the particular considered case.

Noise

- SNR on the measured data : $SNR = \{50; 40; 30; 20; 10\} [dB]$;
- Noise seed : $Noise_Seed = 11$.

Near-Field vs Magnitude Failure Factor ($\nu^{(3)}$)

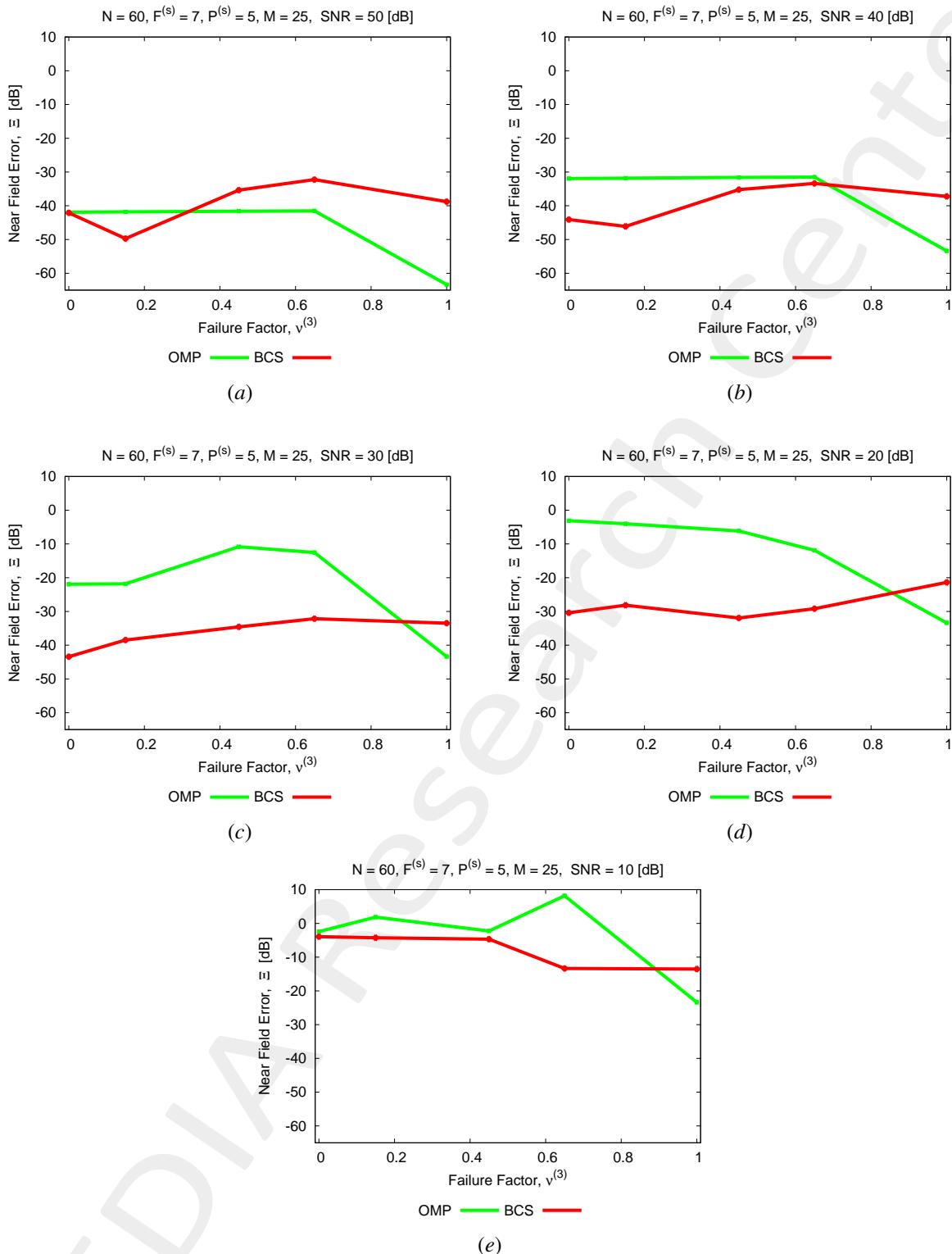


Figure 3: Behaviour of the near-field matching error, Ξ , as a function of the magnitude failure factor affecting the 3rd row of the AUT, $\nu^{(3)}$: (a) $SNR = 50$ [dB], (b) $SNR = 40$ [dB], (c) $SNR = 30$ [dB], (d) $SNR = 20$ [dB], (e) $SNR = 10$ [dB]

Comments

The reported results show that:

-
- the *OMP* solver is quite insensitive to the variation of the magnitude failure factor (the curve is almost flat) except when the magnitude failure is very low (note: higher the value of $\nu^{(s)}$ and more the *AUT* behaves like the *Gold* antenna); indeed, for $\nu^{(3)} \geq 0.65$, it is possible to observe an important error decrease of about 20 [dB] or more. This is true for all the considered *SNR* values;
 - the near-field error of the *BCS* algorithm, even if there are some small variations of the resulting curve, appears to be more independent from the variation of the magnitude failure factor than the *OMP* solver; indeed, whatever the *SNR* value, the error variation is always less than 10 [dB].

1.1.2 AUT with only a phase failure affecting the 3rd row ($\gamma^{(3)} = [-180, -150, -45, 0, 60, 135, 180]$ [Deg]); incremented failure ranges to build the over-complete basis ($\nu^{(s)} \in [0.0, 1.0]$, $F^{(s)} = 7$, $\gamma^{(s)} \in [-\pi, \pi]$, $P^{(s)} = 5$) and M = 25

Parameters

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- Element excitations : $w_n^{(s)} = 1.0 + j0.0$, $n = 1, \dots, N^{(s)}$, $s = 1, \dots, S$;

Antenna Under Test (AUT - With Defects)

1. Failures of the excitation phase of the 3rd row;
 - Phase shift of the elements in the 3rd row ($s = 3$) : $\gamma^{(3)} = [-\pi, -\frac{5}{6}\pi, -\frac{\pi}{4}, 0, \frac{\pi}{3}, \frac{3}{4}\pi, \pi]$ [rad];

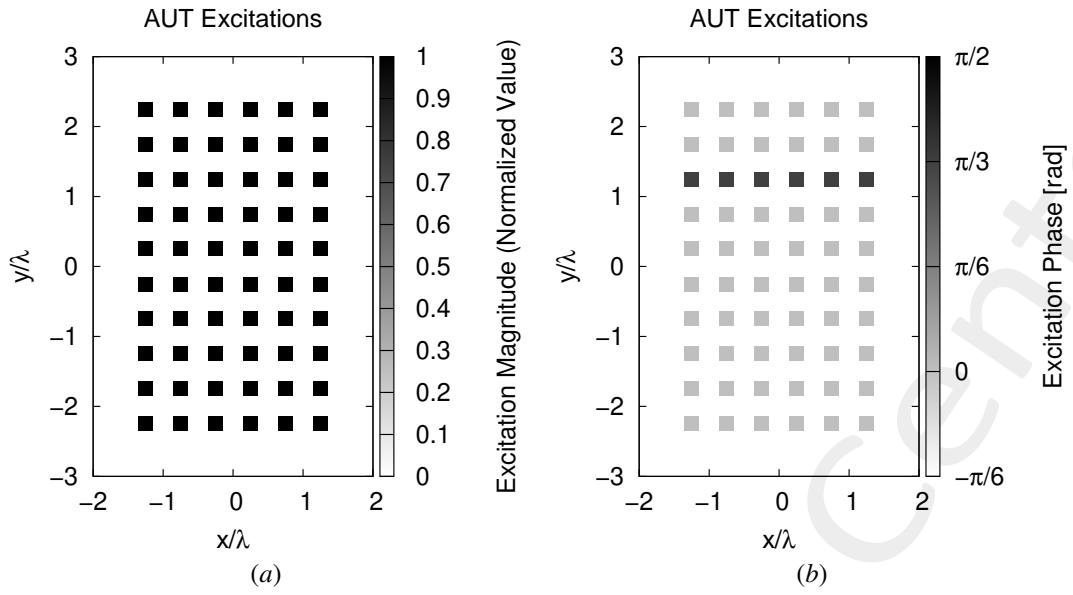


Figure 4: (a) Magnitude of the element excitations in the *AUT* , (b) phase of the element excitations in the *AUT* (e.g. $\gamma^{(3)} = \frac{\pi}{3}$ [rad]).

Measurement Set-Up

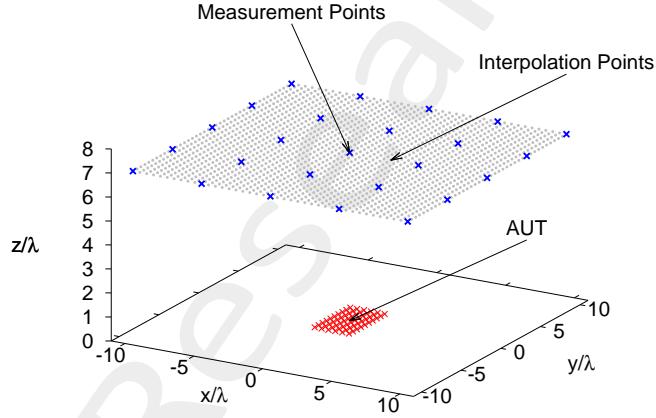


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- Number of points : $M = 25$;
- Measurement step : $\Delta_{x/y}^{meas} = 5 [\lambda]$
- Ratio between number of measurements and total number of elements : $(M/N) = 0.42$;

Measurement-by-Design Technique

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Original (OMP) MbD parameters

- Max. number of iterations of the OMP algorithm : $I = \{1; 2; 3; \dots; 10\}$;
- Selected iteration to report the results: depends on the particular considered case.

Noise

- SNR on the measured data : $SNR = \{50; 40; 30; 20; 10\} [dB]$;
- Noise seed : $Noise_Seed = 63$.

Near-Field vs Phase Shift Factor ($\gamma^{(3)}$)

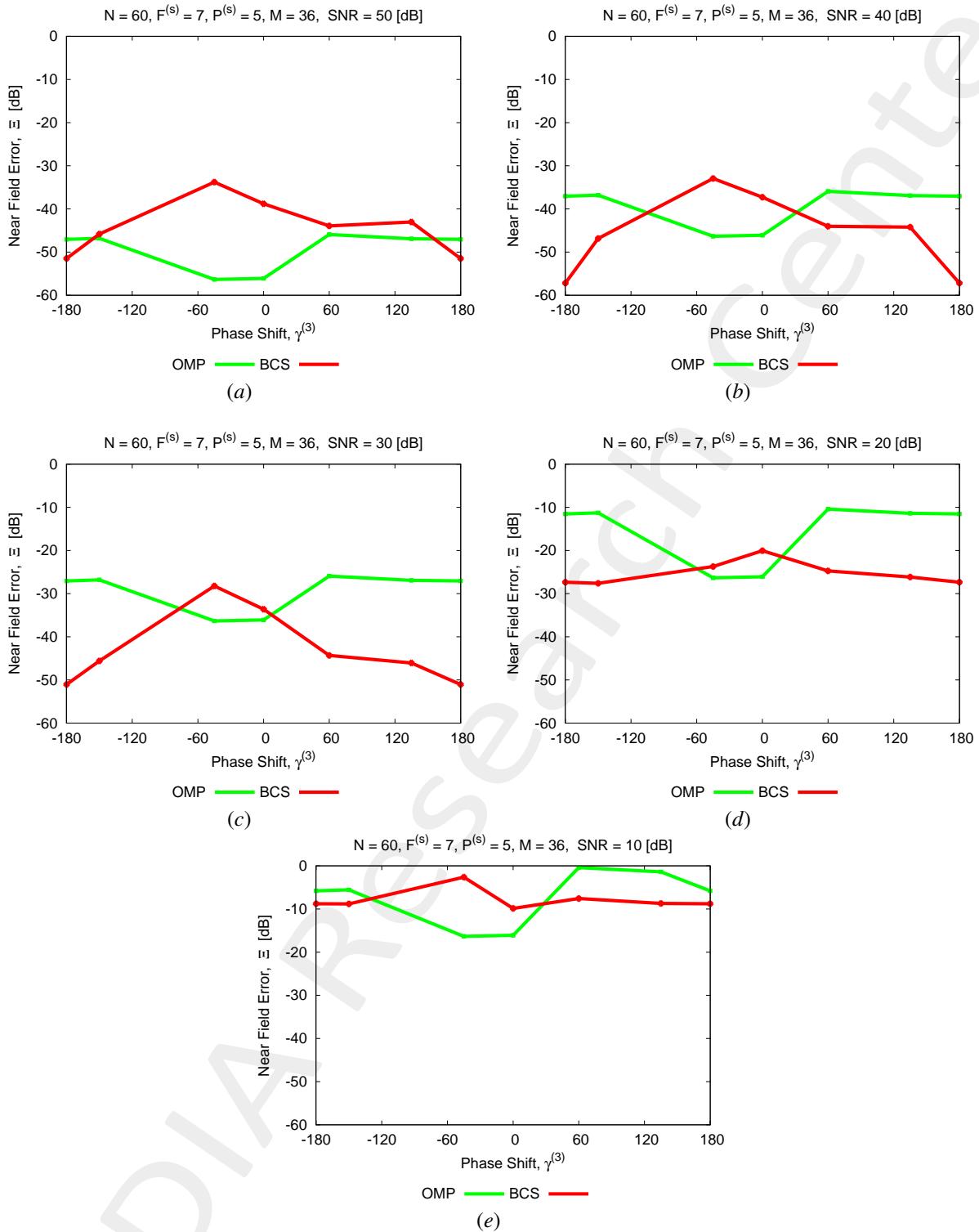


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Comments

The reported results show a slight dependence of the near-field error of both *OMP* and *BCS* solver. The phase shift factors of main interest are those in the middle of the considered range; in particular, the performance of the two algorithms are different because:

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- the *OMP* solver reaches the minimum error for $\gamma^{(3)} = -65$ [Deg] and $\gamma^{(3)} = 0$ [Deg] (the latter case corresponds to the situation in which the AUT has no defects) which results in a resulting curve valley at those points;
 - on the contrary, the *BCS* algorithm, at $\gamma^{(3)} = -65$ [Deg] and $\gamma^{(3)} = 0$ [Deg], obtains its higher error so that its resulting curve presents a peak at those points.

More information on the topics of this document can be found in the following list of references.

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