An Innovative Multi-Frequency PSO-Based Method for the Microwave Imaging of Buried Objects having Different Conductivities

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Abstract

In this work, an innovative particle swarm optimization (*PSO*)-based microwave imaging approach is presented to solve the subsurface inverse scattering problem. The proposed *MF-IMSA-PSO* method integrates a customized *PSO* solver within a multi-scaling technique (i.e., the *IMSA*) in order to limit the ratio between problem unknowns and non-redundant data, mitigating the negative effects of both non-linearity and ill-posedness through the exploitation of progressively *acquired* information about the solution. Moreover, the inversion is performed by considering a multi-frequency (*MF*) solution strategy, by jointly processing several frequency components extracted from the spectrum of the measured data through ground penetrating radar (*GPR*). Some numerical results are shown in order to verify the effectiveness of the developed GPR microwave imaging technique when dealing with objects having a conductivity different from that of the hosting (lossy) soil.

1 Definitions

1.1 Glossary

- SF: Single-Frequency;
- *FH*: Frequency-Hopping;
- *MF*: Multi-Frequency;
- P: Swarm dimension;
- U: Total number of unknowns;
- S: Maximum number of IMSA zooming steps;
- s^{best} : Last performed *IMSA* zooming step $(s^{best} \leq S)$;
- η_{th} : IMSA zooming threshold;
- D_{inv} : Investigation domain;
- D_{obs} : Observation domain;
- L: Side of the investigation domain;
- N: Number of discretization cells in D_{ind} ;
- V: Number of views;
- M: Number of measurement points;
- F: Number of frequencies considered for the inversion;
- $\mathbf{r}^{(v)} = (x^{(v)}, y^{(v)})$: Coordinates of the v-th source $(v = 1, \dots, V)$.
- $\mathbf{r}_m^{(v)} = \left(x_m^{(v)}, y_m^{(v)}\right)$: Coordinates of the *m*-th measurement point for the *v*-th view *v*, $(m = 1, \dots, M)$;
- $\varepsilon_{ra} = \frac{\varepsilon_a}{\varepsilon_0}$: Relative electric permittivity for the upper half-space (y > 0);
- σ_a : Conductivity for the upper half-space (y > 0);
- $\varepsilon_{rb} = \frac{\varepsilon_b}{\varepsilon_0}$: Background relative electric permittivity;
- σ_b : Background conductivity;
- $E_{inc}^{(v)}(\mathbf{r}_n; f)$: Measured internal incident field inside the *n*-th cell, for the *v*-th view at frequency f;
- $\widetilde{E}_{inc}^{(v)}(\mathbf{r}_n; f)$: Computed internal incident field inside the *n*-th cell, for the *v*-th view at frequency f;
- $E_{scatt}^{(v)}\left(\mathbf{r}_{m}^{(v)};f\right)$: Measured external scattered by the *m*-th measurement point, for the *v*-th view at frequency f;
- $\widetilde{E}_{scatt}^{(v)}\left(\mathbf{r}_{m}^{(v)};f\right)$: Measured external scattered by the *m*-th measurement point, for the *v*-th view at frequency f.

1.2 Contrast function

The contrast function at frequency f is defined as

$$\tau\left(\mathbf{r};f\right) = \frac{\varepsilon_{eq}\left(\mathbf{r}\right) - \varepsilon_{eqb}}{\varepsilon_{0}} = \left[\varepsilon_{r}\left(\mathbf{r}\right) - \varepsilon_{rb}\right] + j\left[\frac{\sigma_{b} - \sigma\left(\mathbf{r}\right)}{2\pi f\varepsilon_{0}}\right]$$

where

- $\mathbf{r} = (x, y)$: position vector;
- $\Re \{\tau (\mathbf{r}; f)\} = [\varepsilon_r (\mathbf{r}) \varepsilon_{rb}];$
- $\Im \{\tau (\mathbf{r}; f)\} = \left[\frac{\sigma_b \sigma(\mathbf{r})}{2\pi f \varepsilon_0}\right];$
- $\varepsilon_{eq}(\mathbf{r}) = \varepsilon_0 \varepsilon_r(\mathbf{r}) j \frac{\sigma(\mathbf{r})}{2\pi f};$
- $\varepsilon_{eqb} = \varepsilon_0 \varepsilon_{rb} j \frac{\sigma_b}{2\pi f};$
- ε_r (**r**): relative electric permittivity at position **r**;
- $\sigma(\mathbf{r})$: conductivity at position \mathbf{r} ;

NOTE: we assume that $\varepsilon_r(\mathbf{r})$ and $\sigma(\mathbf{r})$ are **not frequency dependent** (non-dispersive mediums).

1.2.1 Contrast function and reference frequency f_{ref} (MF approaches)

The contrast function at a generic frequency f can be expressed by means of the contrast function computed for a selected reference frequency

$$f = f_{ref} \tag{1}$$

as follows

$$\tau(\mathbf{r}; f) = \Re \left\{ \tau(\mathbf{r}; f_{ref}) \right\} + j \frac{f_{ref}}{f} \Im \left\{ \tau(\mathbf{r}; f_{ref}) \right\}.$$
(2)

This allows to reduce the number of unknowns when dealing with multi-frequency techniques, since we can just consider the contrast function at the reference frequency.

1.3 Cost function & unknowns

1.3.1 Multi-Frequency (MF) approaches

These approaches jointly consider data at F frequencies. The functional minimized by the inversion algorithm is defined as

$$\Phi\left(\mathbf{x}\right) = \Phi_{state}\left(\mathbf{x}\right) + \Phi_{data}\left(\mathbf{x}\right) \tag{3}$$

where $\Phi_{state}(\mathbf{x})$ and $\Phi_{data}(\mathbf{x})$ are respectively the data and state terms of the cost function, defined as

$$\Phi_{state}\left(\mathbf{x}\right) = \frac{\sum_{j=1}^{F} \sum_{v=1}^{V} \sum_{n=1}^{N} \left| E_{inc}^{(v)}\left(\mathbf{r}_{n}; f_{j}\right) - \widetilde{E}_{inc}^{(v)}\left(\mathbf{r}_{n}; f_{j}\right) \right|^{2}}{\sum_{j=1}^{F} \sum_{v=1}^{V} \sum_{n=1}^{N} \left| E_{inc}^{(v)}\left(\mathbf{r}_{n}; f_{j}\right) \right|^{2}}$$
(4)

$$\Phi_{data} = \frac{\sum_{j=1}^{F} \sum_{v=1}^{V} \sum_{m=1}^{M} \left| E_{scatt}^{(v)} \left(\mathbf{r}_{m}^{(v)}; f_{j} \right) - \widetilde{E}_{scatt}^{(v)} \left(\mathbf{r}_{m}^{(v)}; f_{j} \right) \right|^{2}}{\sum_{j=1}^{F} \sum_{v=1}^{V} \sum_{m=1}^{M} \left| E_{scatt}^{(v)} \left(\mathbf{r}_{m}^{(v)}; f_{j} \right) \right|^{2}}$$
(5)

The unknowns of the inversion problem are

$$\mathbf{x} = \left\{ \tau \left(\mathbf{r}; f_{ref} \right); E_{tot}^{(v)} \left(\mathbf{r}_n; f_j \right) \right\} \qquad n = 1, ..., N; v = 1, ..., V; j = 1, ..., F.$$
(6)

The total number of unknowns for MF-based approaches is then given by

$$U_{MF} = 2N\left(1 + VF\right).\tag{7}$$

1.4 Reconstruction errors

The following integral error is defined

$$\Xi_{reg} = \frac{1}{N_{reg}} \sum_{n=1}^{N_{reg}} \frac{|\tau_n^{act} - \tau_n^{rec}|}{|\tau_n^{act} + 1|}$$
(8)

where reg indicates if the error computation covers

- the overall investigation domain $(reg \Rightarrow tot)$,
- the actual scatterer support $(reg \Rightarrow int)$,
- or the background region $(reg \Rightarrow ext)$.

2 Numerical Results: Variation of the Object Conductivity

2.1 *I*-Shaped object ($\varepsilon_{r,obj} = 5.5$)

2.1.1 Parameters

Background

Inhomogeneous and nonmagnetic background composed by two half spaces

- Upper half space (y > 0 air): $\varepsilon_{ra} = 1.0, \sigma_a = 0.0;$
- Lower half space (y < 0 soil): $\varepsilon_{rb} = 4.0, \ \sigma_b = 10^{-3} [\mathrm{S/m}];$

Investigation domain (D_{inv})

- Side: $L_{D_{inv}} = 0.8$ [m];
- Barycenter: $\left(x_{bar}^{D_{inv}}, y_{bar}^{D_{inv}}\right) = (0.00, -0.4) \text{ [m]};$

Time-Domain forward solver (FDTD - GPRMax2D)

- Side of the simulated domain: L = 6 [m];
- Number of cells: $N^{FDTD} = 750 \times 750 = 5.625 \times 10^5$;
- Side of the FDTD cells $l^{FDTD} = 0.008$ [m];
- Simulation time window: $T^{FDTD} = 20 \times 10^{-9}$ [sec];
- Time step: $\Delta t^{FDTD} = 1.89 \times 10^{-11}$ [sec];
- Number of time samples: $N_t^{FDTD} = 1060;$
- Boundary conditions: perfectly matched layer (*PML*);
- Source type: Gaussian mono-cycle (first Gaussian pulse derivative, called "Ricker" in GPRMax2D)
 - Central frequency: $f_0 = 300 \text{ [MHz]};$
 - Source amplitude: A = 1.0 [A];



Figure 1: GPRMax2D excitation signal. (a) Time pulse, (b) normalized frequency spectrum.

Frequency parameters

- Frequency range: $f \in [f_{min}, f_{max}] = [200.0, 600.0] [MHz] (-3 [dB] bandwidth of the Gaussian Monocycle excitation centered at <math>f_0 = 300 [MHz]$);
- Frequency step: $\Delta f = 100 \text{ [MHz]}$ ($F = 5 \text{ frequency steps in } [f_{min}, f_{max}]$);

f [MHz]	$\lambda_a [m]$	$\lambda_b [\mathrm{m}]$	f^* [MHz]
200.0	1.50	0.75	200.5
300.0	1.00	0.50	297.6
400.0	0.75	0.37	401.1
500.0	0.60	0.30	498.1
600.0	0.50	0.25	601.6

Table 1: Considered frequencies and corresponding wavelength in the upper medium (λ_a , free space) and in the lower medium (λ_b , soil). f^* is the nearest frequency sample available from transformed time-domain data, and represents the real frequency considered by the inversion algorithm.

Scatterer

- Type: *I*-Shaped;
- Side: 0.28 [m];
- Electromagnetic properties: $\varepsilon_{r,obj} = 5.5$, $\sigma_{obj} = \{10^{-4}; 5 \times 10^{-4}; 10^{-3}; 5 \times 10^{-3}; 10^{-2}\}$ [S/m];

$\varepsilon_{r,obj}$	$\sigma_{obj} ~[{ m S/m}]$	$\Re\left\{ au ight\}$	$\Im\{\tau\}$
5.5	10^{-4}	1.5	0.040
5.5	5×10^{-4}	1.5	0.022
5.5	10^{-3}	1.5	0.000
5.5	5×10^{-3}	1.5	-0.180
5.5	10^{-2}	1.5	-0.404

Table 2: Real and imaginary parts of the contrast function vs. different values of object conductivity. The imaginary part is computed as $\Im \{\tau\} = \left[\frac{\sigma_b - \sigma_{obj}}{2\pi f \epsilon_0}\right]$ at the central frequency $(f_{cent} = 400 \text{ [MHz]})$.



Measurement setup

- Considered frequency: $f_{min}=200~[{\rm MHz}],\,\lambda_b=0.75~[{\rm m}].^{-1}$
- $\#DoFs = 2ka = \frac{2\pi}{\lambda_b}L\sqrt{2} = \frac{2\pi}{0.75}0.8\sqrt{2} \simeq 9.5;$
- Number of views (sources): V = 10;
 - $-\min\{x_v\} = -0.5 \text{ [m]}, \max\{x_v\} = 0.5 \text{ [m]};$
 - height: $y_v = 0.1 \, [m], \, \forall v = 1, \dots, V;$
- Number of measurement points: M = 9;
 - $-\min\{x_m\} = -0.5 \text{ [m]}, \max\{x_m\} = 0.5 \text{ [m]};$
 - height: $y_m = 0.1 \, [m], \, \forall m = 1, \dots, M;$

¹NOTE: This choice is done in order to keep the number of unknowns lower than 5000.



Figure 3: Location of the measurement points (M = 9) and of the sources (V = 10). Only one source is active for each view.

Inverse solver parameters

• Shared parameters

- Number of unknowns: U = 2N(1 + VF) = 4998;
- Weight of the state term of the functional: 1.0;
- Weight of the data term of the functional: 1.0;
- Weight of the penalty term of the functional: 0.0;
- Convergence threshold: 10^{-10} ;
- Variable ranges:
 - * $\varepsilon_r \in [4.0, 5.8];$
 - $\ast \ \Re\left\{E_{tot}^{int}\right\} \in [-8,8], \ \Im\left\{E_{tot}^{int}\right\} \in [-8,8];$
- Degrees of freedom:

* Considered frequency:
$$f_{min} = 200 \text{ [MHz]}, \lambda_b = 0.75 \text{ [m]};$$

* $\frac{(2ka)^2}{2} = \frac{\left(2 \times \frac{2\pi}{\lambda_b} \times \frac{L\sqrt{2}}{2}\right)^2}{2} = 4\pi^2 \left(\frac{L}{\lambda_b}\right)^2 = 4\pi^2 \left(\frac{0.8}{0.75}\right)^2 \simeq 44.87;$

- Number of cells: $N = 49 = 7 \times 7;$
- Maximum number of IMSA steps: S = 4;
- Side ratio threshold: $\eta_{th} = 0.2;$

• *MF* – *IMSA* – *PSO* parameters

- Maximum number of iterations: I = 20000;
- Swarm dimension: $P = \frac{5}{100} \times U = 250;$
- $-C_1 = C_2 = 2.0;$
- Inertial weight: w = 0.4;

- Velocity clamping: enabled;

- MF IMSA CG parameters
 - Maximum number of iterations: I = 200;

Signal to noise ratio (on $E_{tot}(t)$)

• $SNR = \{50, 40, 30, 20\} [dB] + Noiseless data.$



 $\sigma_{obj} = 10^{-4} [S/m] (\Im \{\tau\} = 0.040) - MF - IMSA - PSO$ vs. MF - IMSA - CG: Final 2.1.2

9 Figure 4: MF - IMSA - PSO vs. MF - IMSA - CG: Retrieved dielectric profiles at the IMSA convergence



2.1.3 $\sigma_{obj} = 5 \times 10^{-4} \text{ [S/m]} (\Im \{\tau\} = 0.022) - MF - IMSA - PSO \text{ vs. } MF - IMSA - CG: Final reconstructions$

Figure 5: MF - IMSA - PSO vs. MF - IMSA - CG: Retrieved dielectric profiles at the IMSA convergence

2.1.4 $\sigma_{obj} = 10^{-3} [S/m] (\Im \{\tau\} = 0.0) - MF - IMSA - PSO$ vs. MF - IMSA - CG: Final reconstructions



Figure 6: MF - IMSA - PSO vs. MF - IMSA - CG: Retrieved dielectric profiles at the IMSA convergence



2.1.5 $\sigma_{obj} = 5 \times 10^{-3} [S/m] (\Im \{\tau\} = -0.180) - MF - IMSA - PSO vs. MF - IMSA - CG: Final reconstructions$

Figure 7: MF - IMSA - PSO vs. MF - IMSA - CG: Retrieved dielectric profiles at the IMSA convergence



2.1.6 $\sigma_{obj} = 10^{-2} [S/m] (\Im \{\tau\} = -0.404) - MF - IMSA - PSO vs. MF - IMSA - CG: Final reconstructions$

Figure 8: MF - IMSA - PSO vs. MF - IMSA - CG: Retrieved dielectric profiles at the IMSA convergence



Figure 9: MF - IMSA - PSO vs. MF - IMSA - CG: Reconstruction errors vs. the object conductivity (σ_{obj}) .



Figure 10: MF - IMSA - PSO vs. MF - IMSA - CG: Reconstruction errors vs. SNR.

3 Conclusions

The reported results indicate that

- The proposed MF-IMSA-PSO imaging technique yields accurate reconstructions also when considering a variation of the conductivity of the buried scatterer;
- On average a significant improvement of the retrieved profiles is obtained with respect to the MF IMSA CG approach, which is based on a deterministic conjugate gradient (CG) solver [5].

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